

Monte Carlo Estimation of PDM-QPSK/OOK and DQPSK/OOK Hybrid Systems Tolerance Against Nonlinear Effects

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Abstract—We compare through Monte Carlo simulations the performance of polarization-division multiplexed quadrature phase-shift keying (PDM-QPSK)/on-off keying (OOK) and differential QPSK (DQPSK)/OOK hybrid systems, taking into account nonlinear phase noise. DQPSK turns out to be better with optimal dispersion-management, while PDM-QPSK proves to be a good candidate for bit rates higher than 40 Gb/s.

Index Terms—Coherent detection, cross-phase modulation (XPM), optical transmission, phase-shift keying.

I. INTRODUCTION

THE Internet traffic growth calls for an increased capacity of dense wavelength division multiplexing (DWDM) optical transmission systems. Advanced modulation formats [1] allow us to design optical systems with a bit rate per channel of 40 Gb/s or higher and good tolerance against linear and nonlinear impairments. The complete substitution of deployed non-return to zero on-off keying (NRZ-OOK) systems is a very expensive option, thus a cost effective solution is to upgrade one or more selected channels, following the market demand. Systems that employ two or more modulation formats and possibly different data rates are commonly referred to as “hybrids”.

Many different formats have been proposed as candidates for the channel upgrade. Phase shaped binary transmission (PSBT) and differential quadrature phase-shift keying (DQPSK) have been studied both numerically [2] and experimentally [3], [4] in hybrid systems. To the authors knowledge, polarization-division multiplexing coherent QPSK (PDM-QPSK) has been only studied once experimentally in a hybrid environment [5]. PSBT and DQPSK require a simpler incoherent receiver, but allow for a maximum bit rate of 40 Gb/s in a DWDM system (channel spacing 50 GHz). On the other hand, PDM-QPSK needs a quite complex coherent receiver and a postreception digital signal processing (DSP) unit, but allows for very high bit rates (80 or 100 Gb/s) on the same grid.

The purpose of this letter is to compare numerically by means of Monte Carlo simulations the performance of PDM-QPSK and DQPSK in a hybrid scenario and try to determine which

one is the most suitable for a system upgrade. We will account for linear and nonlinear system impairments and will highlight pros and cons of the proposed solutions.

II. SIMULATION SETUP

All the simulations were performed using an internally developed optical transmission simulator. We tested a five-channel system on a 50-GHz grid that propagates along a dispersion managed (DM) system composed of 20×100 km spans of Terlight fiber ($D = 8$ ps/nm/km, $\alpha = 0.2$ dB/km, $\gamma = 1.7$ (1/W · km) at 1550 nm) and single mode fiber (SMF, $D = 16$ ps/nm/km, $\alpha = 0.2$ dB/km, $\gamma = 1.3$ (1/W · km) at 1550 nm). All the results refer to the central channel. We separately verified that increasing the number of channels does not cause significant variations of performance [3]. The line was composed of 20 identical spans, each comprising transmission and linear compensating fibers, followed by an amplifier with flat gain and a noise figure of 6 dB. Purely linear pre-/post-compensating fibers were inserted before/after the transmission link.

The even channels were always NRZ-OOK modulated at a bit rate of 10 Gb/s (10 Gbaud). The odd channels were in turn 40 Gb/s DQPSK (20 Gbaud), 40 Gb/s PDM-QPSK (10 Gbaud), or 80 Gb/s PDM-QPSK (20 Gbaud). The OOK channels were modulated using pseudo-random binary sequences (PRBS) with different seeds and length 2^9 , while the DQPSK and PDM-QPSK were modulated using a pseudo-random quaternary sequence (PRQS) of length 4^5 . All channels had the same average power and were synchronous at the input of the pre-fiber. We verified that inserting random delays among channels shows very limited impact thanks to the decorrelation induced by pre-compensating and inline fibers. OOK channels were copolarized with one polarization of coherent channels in PDM-QPSK configuration.

The propagation of signal and amplified spontaneous emission (ASE) noise along the fibers was modeled using a variable step-size split step Fourier method (SSFM), that takes into account all linear and nonlinear effects, except polarization mode dispersion (PMD). The maximum nonlinear phase rotation per step was $3 \cdot 10^{-3}$ rad which we verified to be small enough for the considered system.

The DQPSK channel was received using a second order super-Gaussian optical filter with 3-dB bandwidth $2 \times$ baud rate (20 or 40 GHz), followed by a standard receiver and a fifth-order Bessel filter with 3-dB bandwidth $0.65 \times$ baud rate (6.5 or 13 GHz) [6]. The PDM-QPSK receiver used the same filters and was composed of two coherent mixers (one for each polarization), used to combine the signal with the local

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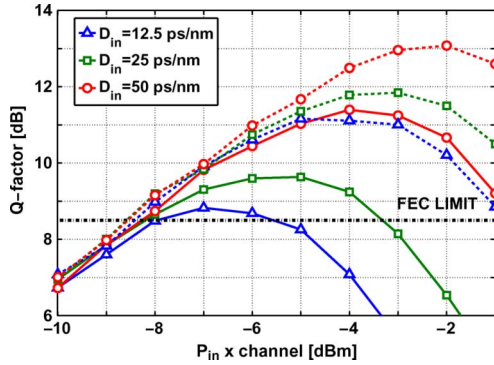


Fig. 1. Q -factor versus P_{in} for DQPSK at 40 Gb/s with 10G OOK neighboring channels. Dashed lines with SMF fiber, solid lines with Teralight.

oscillator, followed by four balanced photodetectors and the DSP unit. The incoming signal is sampled at twice the symbol rate and no quantization is performed. The receiver is the same as described in [7], with no digital dispersion compensation.

III. RESULTS AND DISCUSSION

For each configuration, we measured the Q -factor as a function of the average launched power P_{in} , for three different values of residual dispersion per span, $D_{in} = 12.5, 25$, and 50 ps/nm and using Teralight [8] and SMF. The Q -factor was computed from Monte Carlo simulations of the bit-error rate (BER) in order to account for nonlinear phase noise, which is often neglected [2]. The simulations were stopped when the relative estimation error on BER reached 20% with a Gaussian confidence of 95%, providing in each case at least 100 error counts. The precompensating fiber cumulated dispersion was $-292, -411$, and -649 ps/nm for the three tested D_{in} , respectively, when using Teralight fiber, and $-488, -607$, and -844 ps/nm when using SMF. These values were chosen using the “straight line rule” [9]. The dispersion of the post-compensating fiber was optimized for a QPSK transmission using Karhunen–Loève method [6], assuming white noise. Typically, the residual total dispersion was close to zero and in every tested case it was within the range $[-40; 40]$ ps/nm.

In such systems, the main nonlinear impairment was found to be the cross-phase modulation (XPM) due to 10G OOK channels on 40G/80G channels [3]. Though not reported here, we ran single channel simulations for DQPSK and 80G PDM-QPSK configurations with Teralight fiber. We verified that cross-channel effects are dominant. Among these, four-wave mixing is negligible because of the sizable inline dispersion, while switching cross-polarization effects in simulation of PDM-QPSK systems did improve the Q -factor by no more than 0.7 dB over the shown input power range. Hence, we conclude that XPM is by far the dominant nonlinear effect.

Figs. 1–3 sketch the measured Q -factor versus launched power for the configurations under investigation. The best performance is always obtained with SMF fiber which is not surprising since increasing the dispersion is known to reduce XPM. Also, the use of SMF reduces the impact of D_{in} (the Q -factor curves for SMF are closer than for Teralight), thus relaxing the dispersion mapping constraints. From a comparison of Fig. 1 and Fig. 2, we note that the best Q -factor is better for the DQPSK case (~ 0.8 dB with Teralight and ~ 1.5 dB with SMF). The 40G PDM-QPSK shows an enhanced

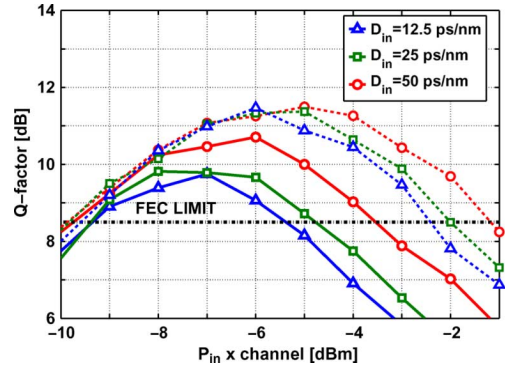


Fig. 2. Q -factor versus P_{in} for PDM-QPSK at 40 Gb/s with 10G OOK neighboring channels. Dashed lines with SMF fiber, solid lines with Teralight.

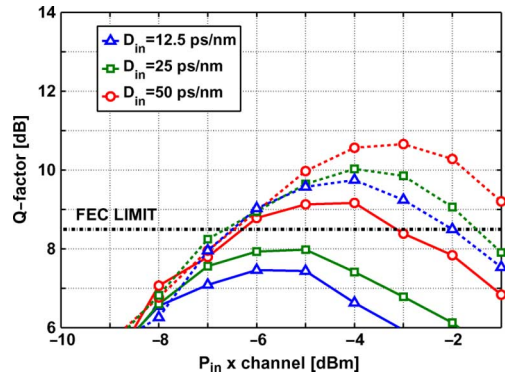


Fig. 3. Q -factor versus P_{in} for PDM-QPSK at 80 Gb/s with 10G OOK neighboring channels. Dashed lines with SMF fiber, solid lines with Teralight.

TABLE I
IMPROVEMENT OF BEST Q -FACTOR [in decibels] WHEN
USING 100 GHz SPACING ON TERALIGHT FIBER

	DQPSK 40G	PDM-QPSK 40G	PDM-QPSK 80G
$D_{in} = 12.5$ ps/nm	1.96	0.78	0.33
$D_{in} = 25$ ps/nm	2.24	1.17	0.24
$D_{in} = 50$ ps/nm	2.12	0.83	0.24

optical signal-to-noise ratio (OSNR) sensitivity in the linear region on the left of the maximum Q -factor. On the other hand, DQPSK shows a superior nonlinear threshold (NLT, defined as P_{in} per channel at best Q -factor). Fig. 3 shows that 80G PDM-QPSK suffers from a reduced OSNR sensitivity due to its higher bit rate (~ 3 dB in Q -factor at $P_{in} = -8$ dBm) and for low values of D_{in} the nonlinear penalties are enhanced, compared to 40G PDM-QPSK. However, the NLT is better than 40G PDM-QPSK (~ 2 dB with both fibers). Over such a long distance this configuration cannot offer adequate performance. However, with a careful DM (higher values of D_{in}) and on shorter links ($1000 \div 1500$ km), PDM-QPSK could be a good candidate for 80G or even 100G channel upgrades.

We also investigated the effect of increasing the channel spacing from 50 to 100 GHz on a Teralight fiber. Table I reports the improvement of the best Q -factor for the considered configurations. As expected, the Q -factor is always better at 100 GHz, but this effect is more evident for DQPSK (~ 2 dB) than for 40/80G PDM-QPSK (~ 1 and ~ 0.3 dB, respectively). This is probably due to the fact that increasing the channel spacing does not reduce nonlinear depolarization [10].

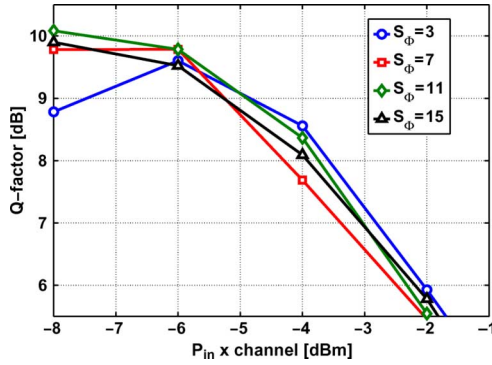


Fig. 4. Q -factor versus P_{in} of PDM-QPSK 40 GB/s+OOK, $D_{in} = 25$ ps/nm on Teralight fiber for four different values of S_{Φ} .

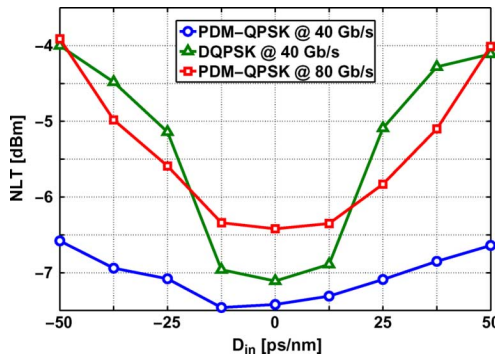


Fig. 5. NLT versus D_{in} of the three examined configurations on Teralight fiber.

A possible solution to improve PDM-QPSK robustness against SPM and XPM is to improve the phase estimation algorithm. In our simulations, we used a Viterbi&Viterbi [7]. The number of samples used to average the phase (S_{Φ}) was seven. It is a well-known fact that the value of S_{Φ} has an influence on the quality of the estimation process; high values of S_{Φ} are better in ASE-limited systems, while smaller values are better when nonlinear effects arise. Fig. 4 shows the performance of the 40G PDM-QPSK+OOK system, using $D_{in} = 25$ ps/nm and Teralight fiber, for different values of S_{Φ} . $S_{\Phi} = 7$ is near the optimum for low values of P_{in} , but gives a penalty of almost 1 dB for $P_{in} = -4$ dBm. Using smaller values of S_{Φ} (es. 3) could thus reduce the impact of XPM, but could be insufficient to provide a decisive improvement. On the other hand, the DSP unit allows for a suboptimal DM design. Comparing Figs. 1 and 2 for $D_{in} = 12.5$ ps/nm with Teralight fiber, we see that the Q -factor of PDM-QPSK is decreased by 2 dB w.r.t. $D_{in} = 50$ ps/nm, while the Q -factor of DQPSK is 4.5 dB smaller. The enhanced tolerance of PDM-QPSK is due to the constant modulus algorithm (CMA) used at the receiver that acts as a generic adaptive equalizer, thus relaxing the impact of DM [7].

Finally, Fig. 5 shows the NLT as a function of D_{in} using Teralight fiber. These curves are obtained by interpolating the available values of Q -factor versus P_{in} . The NLT always grows for increasing values of $|D_{in}|$, because residual inline dispersion reduces the dominant XPM impairment. The NLT of 40G

PDM-QPSK is the lowest but its value is little dependent on D_{in} (1 dB). On the other hand, the NLT for 40G DQPSK and 80G PDM-QPSK is higher, but it decreases by more than 3 and 2.5 dBm, respectively, at $D_{in} = 0$.

IV. CONCLUSION

We provided a numerical Monte Carlo comparison between 40G DQPSK, 40G PDM-QPSK, and 80G PDM-QPSK, three of the best candidates for the deployment of hybrid systems, assessing their tolerances against non-linear effects.

PDM-QPSK is a very promising format that will enable us to deliver bit rates up to 100 Gb/s per channel. However, in order to ensure enhanced tolerance against cross-channel nonlinear impairments and thus extend the reach, the phase should be carefully estimated. At 40 Gb/s, PDM-QPSK is strongly affected by XPM while, with proper DM design, DQPSK has proven to be the best choice among the considered formats for this bit rate in ultra-long-haul systems.

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